#### **General Description**

The LTA805x family is true single-supply voltage feedback operational amplifiers feature high speed performance with 200 MHz of small signal bandwidth and 160 V/ $\mu$ s slew rate. The products are specified for 3.3 V and 5 V supplies, input common mode voltage range extends to 0.1 V below V<sub>s</sub> and 1 V from V<sub>s+</sub>, and output voltage range extends to power rail, allowing wide dynamic range especially desirable in low voltage applications. The LTA805x also offer excellent signal quality of low distortion and fast settling time (13.5 ns to 0.1%), which make them ideal as buffers to single-supply ADCs.

Operating on supplies from +3.3 V to +6.6 V and dual supplies up to ±3.3 V, the LTA805x are ideal for a wide range of applications, from battery-operated systems with large bandwidth requirements to high speed systems where component density requires lower power dissipation. The single version LTA8051 device is available in micro-size S0T23-5L and S0IC-8L packages. The dual LTA8052 device is offered in MSOP-8L and S0IC-8L packages. The quad LTA8054 device is offered in S0IC-14L and TSSOP-14L packages.

#### **Features and Benefits**

- High Speed and Fast Settling on 5 V
- 200 MHz, -3 dB bandwidth (G = +1)
- 160 V/µs slew rate
- 13.5 ns settling time to 0.1%
- Fully specified at 3.3 V and 5 V Supplies
- Input Common Mode Voltage 0.1 V Beyond V<sub>s−</sub>, 1 V from V<sub>s+</sub>
- Output Short Circuit Current 120 mA
- Operating Temperature Range -40°C to +125°C

### **Applications**

- Photodiode Amplification
- Video Buffer
- Active Filters
- Driving A/D Converters
- Motor Phase Current Sense
- Portable Equipment
- Battery-Powered Instrumentation



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## Ordering Information<sup>(1)</sup>

Part Number	Package Type	Package Size	Package Quantity	ECO Class <sup>(2)</sup>	Mark Code <sup>(3)</sup>
LTA8051XT5/R6	S0T23-5L	2.92 mm * 1.60 mm	Tape and Reel, 3 000	Green (RoHS & no Sb/Br)	W57
LTA8051XS8/R8	SOIC-8L	4.90 mm * 3.92 mm	Tape and Reel, 4 000	Green (RoHS & no Sb/Br)	W8051
LTA8052XS8/R8	SOIC-8L	4.90 mm * 3.92 mm	Tape and Reel, 4 000	Green (RoHS & no Sb/Br)	W8052
LTA8052XV8/R6	MSOP-8L	3.00 mm * 3.00 mm	Tape and Reel, 4 000	Green (RoHS & no Sb/Br)	W8052
LTA8054XS14/R5*	SOIC-14L	8.73 mm * 3.95 mm	Tape and Reel, 2 500	Green (RoHS & no Sb/Br)	W8054
LTA8054XT14/R6*	TSSOP-14L	4.96 mm * 4.40 mm	Tape and Reel, 3 000	Green (RoHS & no Sb/Br)	W8054

\* Preview Status (Not for MP stage, pls contact with us if you have request)

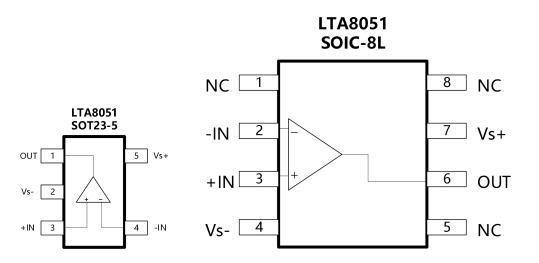
(1) Please contact to your Linearin representative for the latest availability information and product content details.

(2) Eco Class - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS & Halogen Free).

(3) There may be multiple device markings, a varied marking character of "x", or additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

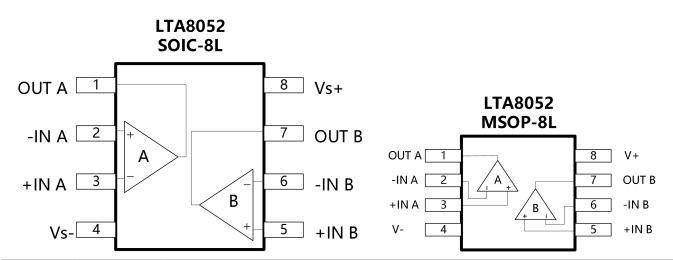


# Pin Configuration (Top View)



PIN Name	S0T23-5L	SOIC-8L	Description
OUT	1	6	Amplifier output.
V <sub>s-</sub>	2	4	Negative power supply. It is normally tied to ground. It can also be tied to a voltage other than ground as long as the voltage between $V_{S+}$ and $V_{S-}$ is from 3.3 V to 6.6 V.
+IN	3	3	Non-inverting input of the amplifier. The voltage range is from (Vs- – 0.1 V) to (Vs+ – 1 V).
-IN	4	2	Inverting input of the amplifier. This pin has the same voltage range as –IN.
Vs+	5	7	Positive power supply. The voltage is from 3.3 V to 6.6 V. Split supplies are possible as long as the voltage between $V_{s+}$ and $V_{s-}$ is from 3.3 V to 6.6 V.
NC	-	1, 5, 8	No Connection

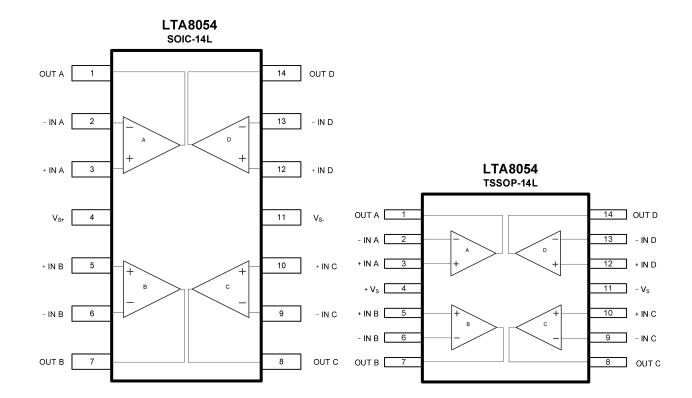




PIN Name	SOIC-8L / MSOP-8L	Description
OUT A	1	Amplifier A output.
-IN A	2	Inverting input A of the amplifier. The voltage range is from (Vs- – 0.1 V) to (Vs+ – 1 V).
+IN A	3	Non-inverting input of the amplifier. This pin has the same voltage range as –IN A.
V <sub>s-</sub>	4	Negative power supply. It is normally tied to ground. It can also be tied to a voltage other than ground as long as the voltage between $V_{S+}$ and $V_{S-}$ is from 3.3 V to 6.6 V.
+IN B	5	Non-inverting input of the amplifier. This pin has the same voltage range as –IN B.
-IN B	6	Inverting input B of the amplifier. The voltage range is from (V $_{S-}$ – 0.1 V) to (V $_{S+}$ – 1 V).
OUT B	7	Amplifier B output.
Vs+	8	Positive power supply. The voltage is from 3.3 V to 6.6 V. Split supplies are possible as long as the voltage between V <sub>S+</sub> and V <sub>S-</sub> is from 3.3 V to 6.6 V.

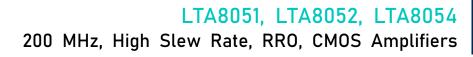


LTA8051, LTA8052, LTA8054 200 MHz, High Slew Rate, RRO, CMOS Amplifiers



PIN Name	SOIC-14L / TSSOP-14L	Description
OUT A	1	Amplifier A output.
-IN A	2	Inverting input A of the amplifier. The voltage range is from (Vs- – 0.1 V) to (Vs+ – 1 V).
+IN A	3	Non-inverting input of the amplifier. This pin has the same voltage range as –IN A.
V <sub>s+</sub>	4	Positive power supply. The voltage is from 3.3 V to 6.6 V. Split supplies are possible as long as the voltage between V <sub>S+</sub> and V <sub>S-</sub> is from 3.3 V to 6.6 V.
+IN B	5	Non-inverting input of the amplifier. This pin has the same voltage range as –IN B.
-IN B	6	Inverting input B of the amplifier. The voltage range is from (Vs- – 0.1 V) to (Vs+ – 1 V).
OUT B	7	Amplifier B output.
OUT C	8	Amplifier C output.
-IN C	9	Inverting input C of the amplifier. The voltage range is from ( $V_{S-}$ – 0.1 V) to ( $V_{S+}$ – 1 V).
+IN C	10	Non-inverting input of the amplifier. This pin has the same voltage range as –IN C.
V <sub>s-</sub>	11	Negative power supply. It is normally tied to ground. It can also be tied to a voltage other than ground as long as the voltage between $V_{S^+}$ and $V_{S^-}$ is from 3.3 V to 6.6 V.
+IN D	12	Non-inverting input of the amplifier. This pin has the same voltage range as –IN D.
-IN D	13	Inverting input D of the amplifier. The voltage range is from (Vs- – 0.1 V) to (Vs+ – 1 V).
OUT D	14	Amplifier D output.





## **Limiting Value**

In accordance with the Absolute Maximum Ratin System (IEC60134).

Parameter	Absolute Maximum Rating
Supply Voltage, $V_{\text{S+}}$ to $V_{\text{S-}}$	10 V
Signal Input Terminals: Voltage	V <sub>S-</sub> – 0.5 V to V <sub>S+</sub> + 0.5 V
Signal Input Terminals: Current	± 10 mA
Output Short-Circuit	Continuous
Storage Temperature Range, T <sub>stg</sub>	–65 °C to +150 °C
Junction Temperature, T	150 °C
Lead Temperature Range (Soldering 10 sec)	260 °C

#### **ESD Ratings**

Parameter	Level	UNIT
Human body model (HBM), per ESDA/JEDEC JS-001-2023 $^{(1)}$	$\pm$ 8 000	V
Charged device model (CDM), per JESD22-A115C-2010	$\pm$ 1 000	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. Manufacturing with less than 500-V HBM is possible if necessary precautions are taken.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Manufacturing with less than 250-V CDM is possible if necessary precautions are taken.

## **Thermal Information**

Thermal Metric		Package	Level	Unit
Αιθ		S0T23-5L	190	
		SOIC-8L	125	
	Package Thermal Resistance	MSOP-8L	216	°C/W
		TSSOP-14L	112	
		SOIC-14L	115	



## **5 V Electrical Characteristics**

Unless otherwise noted, V <sub>S</sub> = ±2.5 V, V <sub>cm</sub> = 0V, A <sub>V</sub> = +1, R <sub>F</sub> = 20 $\Omega$ ; A <sub>V</sub> $\ge$ +2, R <sub>F</sub> = 470 $\Omega$ and R <sub>L</sub> = 100 $\Omega$ .								
Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit		
DYNAMIC PERFORMANCE								
		G= +1, $V_{OUT}$ = 0.2 $V_{P-P}$ , $R_F$ =20 $\Omega$ , $R_L$ = 150 $\Omega$		200				
-3dB Small-Signal		G= +1, V <sub>OUT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =20 Ω, R <sub>L</sub> = 1 kΩ		240		MHz		
Bandwidth	f <sub>-3dB</sub>	G= +2, $V_{OUT}$ = 0.2 $V_{P-P}$ , $R_F$ =470 $\Omega$ , $R_L$ = 150		85				
		Ω		00				
		G= +2, V <sub>0UT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =470 Ω, R <sub>L</sub> = 1 kΩ		110				
Gain-Bandwidth Product	GBP	G= +10, R <sub>L</sub> = 150 Ω		12		MHz		
		G= +10, R∟ = 1 kΩ		12.5				
Bandwidth for 0.1dB Flatness	f <sub>0.1dB</sub>	G = +2, V <sub>OUT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =470 Ω, R <sub>L</sub> = 150 Ω		8		MHz		
Slew Rate	SR	$G = +1, V_{IN} = 2 V_{P-P}$		160/180		V/µs		
Rise Time	Tr	G = +1, V <sub>IN</sub> = 0.2 V Step		2.2		ns		
Fall Time	T <sub>f</sub>	G = +1, V <sub>IN</sub> = 0.2 V Step		1.8		ns		
Settling Time to 0.1%	T₅	G= -1, R <sub>F</sub> = 402 Ω, V <sub>OUT</sub> = 2 V Step		13.5		ns		
NOISE and DISTORTION PERI	ORMANCI	Ε		1				
Input Voltage Noise Density	en	$f \ge 1 MHz$		1		nV/√Hz		
Differential Gain	DG	G = +2, R <sub>L</sub> = 150 Ω		0.05		%		
Differential Phase	DP	G = +2, R <sub>L</sub> = 150 Ω		0.05		0		
DC PERFORMANCE	I	, _	I					
Input Offset Voltage	Vos			±1.8	±15	mV		
Input Offset Voltage vs	dV <sub>os</sub> /							
Temperature	dT	T <sub>A</sub> = -40°C to 125°C		4		μV/°C		
Input Bias Current	IB			0.5		pА		
Input Offset Current	l <sub>os</sub>			0.2		pА		
Open-loop voltage gain	Avol	$R_L$ = 100 Ω, V <sub>0</sub> = 3.3 V <sub>p-p</sub>		80	100	dB		
INPUT CHARACTERISTICS	•		•	•				
Input Common Mode	V <sub>CM</sub>		V <sub>S-</sub>		V <sub>s+</sub> -1	v		
Voltage Range	♥СМ		-0.1		V S+ - I	v		
Common Mode Rejection Rate	CMRR	G = +100, V <sub>CM</sub> = -2.2 V to +1 V	60	80	108	dB		
OUTPUT CHARACTERISTICS		1				1		
		No Load	V <sub>S+</sub> -	V <sub>S+</sub> -				
High Output Voltage Swing	V <sub>он</sub>		80	40		mV		
		R <sub>L</sub> = 100 Ω	V <sub>s+</sub> - 450	V <sub>s+</sub> - 350				
				Vs-	V <sub>S-</sub>			
Low Output Voltage Swing		No Load		+5	+20	mV		
Low Output voltage Swing	V <sub>OL</sub>	R <sub>L</sub> = 100 Ω		V <sub>S-</sub>	V <sub>S-</sub>			
				+100	+250			
Short-circuit Current	Isource	Open loop, Vin = ± 200 mV		120		mA		
	l <sub>sink</sub>			140				
POWER SUPPLY	1			1		1		
Operating Supply Voltage	Vs	T <sub>A</sub> = -40 to +125 °C	3.3		6.6	V		
Quiescent Current (Per amplifier)	la			6.7	8	mA		



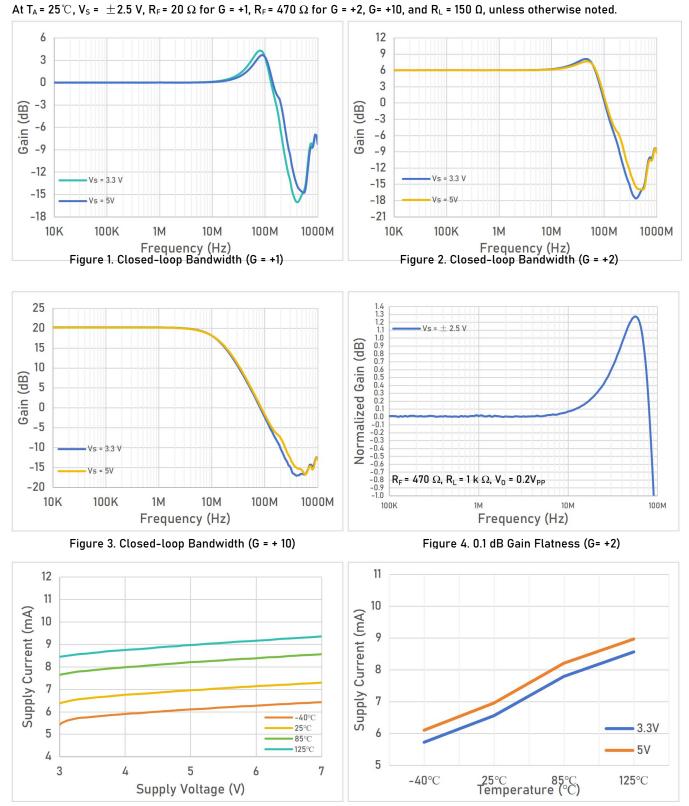
## 3.3 V Electrical Characteristics

Unless otherwise noted,	$V_{c} = \pm 1.65 \text{ V} \text{ V}_{cm} =$	= ΛV Δ <sub>V</sub> = +1 R <sub>F</sub> = 20 O	• Δ <sub>V</sub> > +2 R <sub>c</sub> = 470 O a	nd R <sub>1</sub> = 100 O
onicess other wise noted,	VS 1.05 V, Vcm -	- 08, AV - 11, IVF - 20 32	, Av <u>~</u> · Z, N⊧ - 470 32 a	$\Pi u \Pi \chi_{L} = 100 \ sz.$

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit	
DYNAMIC PERFORMANCE		1					
		G= +1, V <sub>OUT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =20 Ω, R <sub>L</sub> = 150 Ω		160			
-3dB Small-Signal		G= +1, V <sub>OUT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =20 Ω, R <sub>L</sub> = 1 kΩ		200		– MHz	
Bandwidth	f₋ <sub>3dB</sub>	G= +2, V <sub>0UT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =470 Ω, R <sub>L</sub> = 150 Ω		80			
		G= +2, V <sub>0UT</sub> = 0.2 V <sub>P-P</sub> , R <sub>F</sub> =470 Ω, R <sub>L</sub> = 1 kΩ	105				
		G= +10, R <sub>L</sub> = 150 Ω		12			
Gain-Bandwidth Product	GBP	G= +10, R <sub>L</sub> = 1 kΩ		12.5		MHz	
Bandwidth for 0.1dB Flatness	f <sub>0.1dB</sub>	G = +2, $V_{OUT}$ = 0.2 $V_{P-P}$ , $R_F$ =470 $\Omega$ , $R_L$ = 150 $\Omega$		7		MHz	
Slew Rate	SR	G = +1, V <sub>IN</sub> = 2 V <sub>P-P</sub>		150/170		V/µs	
Rise Time	Tr	G = +1, V <sub>IN</sub> = 0.2 V Step		4		ns	
Fall Time	T <sub>f</sub>	G = +1, V <sub>IN</sub> = 0.2 V Step		4		ns	
Settling Time to 0.1%	T₅	G= -1, R <sub>F</sub> = 402 Ω, V <sub>0UT</sub> = 2 V Step		16.5		ns	
NOISE and DISTORTION PER	FORMANCE						
Input Voltage Noise Density	en	$f \ge 1 MHz$		1		nV/√Hz	
DC PERFORMANCE							
Input Offset Voltage	Vos			±1.5	±15	mV	
Input Offset Voltage vs Temperature	dV₀₅ / dT	T <sub>A</sub> = -40°C to 125°C		3		µV/°C	
Input Bias Current	IB			0.6		pА	
Input Offset Current	los			0.6		pА	
Open-loop voltage gain	A <sub>VOL</sub>	$R_{L}$ = 100 $\Omega$ , $V_{o}$ = 1.8 $V_{p\text{-}p}$		70	80	dB	
INPUT CHARACTERISTICS						_	
Input Common Mode Voltage Range	V <sub>CM</sub>		V <sub>s-</sub> -0.1		V <sub>S+</sub> -1	V	
Common Mode Rejection Rate	CMRR	G = +100, V <sub>CM</sub> = -1.65 V to +0.35 V	55	60	90	dB	
OUTPUT CHARACTERISTICS							
High Output Voltage Swing	V <sub>он</sub>	No Load	V <sub>s+</sub> - 50	V <sub>s+</sub> - 30		- mV	
ingi output voltage owing	▼UH	R <sub>L</sub> = 100 Ω	V <sub>s+</sub> - 350	V <sub>s+</sub> - 300			
		No Load		V <sub>s-</sub> +3	V <sub>s-</sub> +5		
Low Output Voltage Swing	V <sub>OL</sub>	R <sub>L</sub> = 100 Ω		V <sub>s-</sub> +180	V <sub>s-</sub> +200	— mV	
Short-circuit Current	I <sub>source</sub>	Open loop, Vin = ± 200 mV		70		mA	
	l <sub>sink</sub>			90			
POWER SUPPLY							
Operating Supply Voltage	Vs	T <sub>A</sub> = -40 to +125 °C	3.3		6.6	V	
Quiescent Current (Per amplifier)	la			6.3	7	mA	

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# **Typical Characteristics**



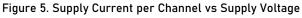


Figure 6. Supply Current per Channel vs Temperature

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## **Typical Characteristics (Cont.)**

At T<sub>A</sub> = 25  $^{\circ}$ C, V<sub>S</sub> = ±2.5 V, R<sub>F</sub> = 20  $\Omega$  for G = +1, R<sub>F</sub> = 470  $\Omega$  for G = +2, G= +10, and R<sub>L</sub> = 150  $\Omega$ , unless otherwise noted.

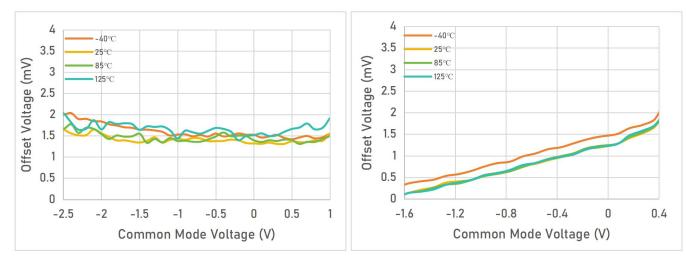


Figure 7. Offset Voltage vs Common Mode Voltage (Vs = ± 2.5V) Figure 8. Offset Voltage vs Common Mode Voltage (Vs = ± 1.65 V)

2.5 2

1.5

0.5

-0.5

-2.5

0

20

40

 $V_{s}$  = ±2.5 V, G = +1, R<sub>L</sub> = 100  $\Omega$ 

60

Figure 10. Output Voltage vs Output Current ( $V_s = \pm 2.5 V$ )

-1 -1.5 -2

1

0

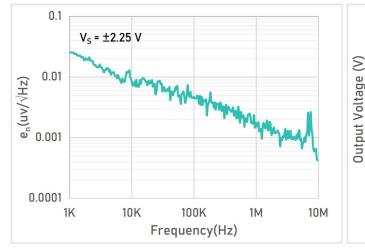
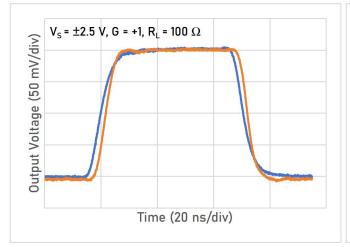


Figure 9. Input Voltage Noise vs Frequency





80

Output Current (mA)

100

120

140

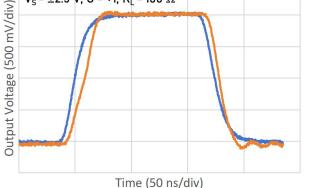


Figure 12. Non-Inverting Large-Signal Step Response



40°C 25°C

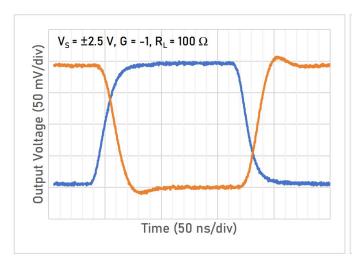
85°C 125°C

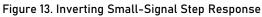
 $V_{s}$  = ±2.5 V, G = -1, R<sub>L</sub> = 100  $\Omega$ 

### **Typical Characteristics (Cont.)**

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At T<sub>A</sub> = 25  $^{\circ}$ C, V<sub>S</sub> = ±2.5 V, R<sub>F</sub> = 20  $\Omega$  for G = +1, R<sub>F</sub> = 470  $\Omega$  for G = +2, G= +10, and R<sub>L</sub> = 150  $\Omega$ , unless otherwise noted.





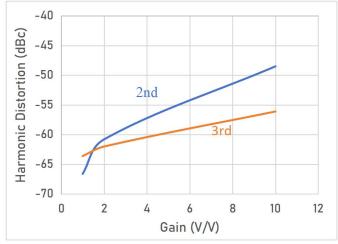


Figure 15. Harmonic Distortion vs Gain  $Vs=\pm 2.5V, f = 1MHz, V_o = 2V_{pp}$ 

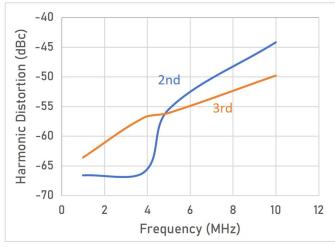


Figure 17. Harmonic Distortion vs Frequency Vs=±2.5V, G = +1, V<sub>o</sub> = 2V<sub>pp</sub>

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Output Voltage (500 mV/div) Time (50 ns/div) Figure 14. Inverting Large-Signal Step Response

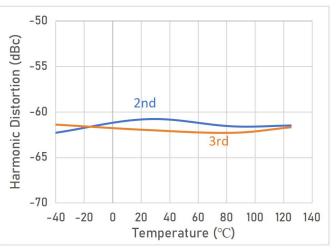


Figure 16. Harmonic Distortion vs Temperature  $Vs=\pm 2.5V, f = 1MHz, V_{o} = 2V_{pp}$ 



#### **Detailed Description**

The LTA805x is a family of High speed, high slew rate, rail-to-rail output operational amplifiers specifically designed for high-speed applications. These devices operate from 3.3 V to 6.6 V at the temperature range of -40 °C to +125 °C, are unity-gain stable, and suitable for a wide range of general-purpose applications. The output stage is capable of driving heavy loads with 130 mA linear output current. The input common-mode voltage range includes from  $V_{s_-}$ -0.1 V to  $V_{s_+}$ -1 V, and allows the LTA805x family to be used in virtually any single supply application. Rail-to-rail output swing significantly increases dynamic range, especially in low-supply applications, and makes them ideal for driving sampling analog-to-digital converters (ADCs).

The LTA805x features 200 MHz bandwidth and 160 V/ $\mu$ s slew rate, providing good ac performance at same time. DC applications are also well served with low input bias current, and an input offset voltage of ±1.8 mV typically. The typical offset voltage drift is 4  $\mu$ V/°C, over the full temperature range the input offset voltage changes only 660  $\mu$ V.

#### **Operating Voltage**

The LTA805x family is optimized for operation at voltages as low as +3.3 V (±1.65 V) and up to +6.6 V (±3.3 V). In addition, many specifications apply from -40 °C to +125 °C. Parameters that vary significantly with operating voltages or temperature are illustrated in the Typical Characteristics graphs.

#### Rail to rail output

Designed as a high speed, high slew rate, low-noise operational amplifier, the LTA805x delivers a robust output drive capability. A class AB output stage with common-source transistors is used to achieve full rail-to-rail output swing capability. For resistive loads up to 100  $\Omega$ , the output swings typically to within 350-mV of 5 V supply rail. Different load conditions change the ability of the amplifier to swing close to the rails. For in open load, the output swings typically to within 40 mV of the positive supply rail and within 5 mV of the negative supply rail.

#### Capacitive load and stability

As with most amplifiers, driving larger capacitive loads than specified may cause excessive overshoot and ringing, or even oscillation. A heavy capacitive load reduces the phase margin and causes the amplifier frequency response to peak. Peaking corresponds to overshooting or ringing in the time domain. Therefore, it is recommended that external compensation be used if the LTA805x op-amps must drive a heavy capacitive load. This compensation is particularly important in the unity-gain configuration, which is the worst case for stability.

A quick and easy way to stabilize the op-amp for capacitive load drive is by adding a series resistor, R<sub>ISO</sub>, between the amplifier output terminal and the load capacitance, as shown in Figure 1. R<sub>ISO</sub> isolates the amplifier output and feedback network from the capacitive load. The bigger the R<sub>ISO</sub> resistor value, the more stable V<sub>OUT</sub> will be. Note that this method results in a loss of gain accuracy because R<sub>ISO</sub> forms a voltage divider with the R<sub>L</sub>.



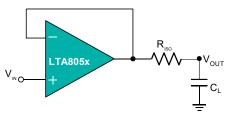


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability. The R<sub>F</sub> provides the DC accuracy by connecting the inverting signal with the output.

The  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

For no-buffer configuration, there are two other ways to increase the phase margin: (a) by increasing the amplifier's gain, or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

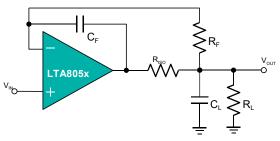


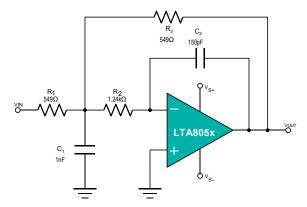
Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

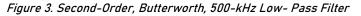


## **Typical Application Circuit**

### **Active filter**

The LTA805x family is well-suited for active filter applications that require a wide bandwidth, fast slew rate, single-supply operational amplifier. Figure 3 shows a 500 kHz, second-order, low-pass filter using the multiple-feedback (MFB) topology. The components have been selected to provide a maximally-flat Butterworth response. Beyond the cut-off frequency, roll-off is -40 dB/dec. The Butterworth response is ideal for applications that require predictable gain characteristics, such as the anti-aliasing filter used in front of an ADC.





One point to observe when considering the MFB filter is that the output is inverted, relative to the input. If this inversion is not required, or not desired, a non- inverting output can be achieved through one of these options:

- 1. adding an inverting amplifier ;
- 2. adding an additional second-order MFB stage;
- 3. using a non-inverting filter topology, such as the Sallen-Key (shown in Figure 4).

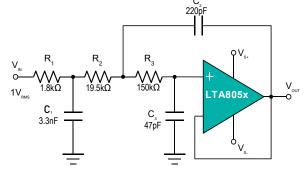


Figure 4. Configured as a Three-Pole, 20-kHz, Sallen- Key Filter

### **Differential amplifier**

The circuit shown in Figure 5 performs the difference function. If the resistors ratios are equal  $R_4/R_3 = R_2/R_1$ , then:  $V_{OUT} = (V_p - V_n) \times R_2/R_1 + V_{REF}$ 

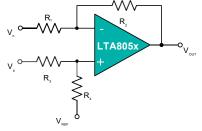
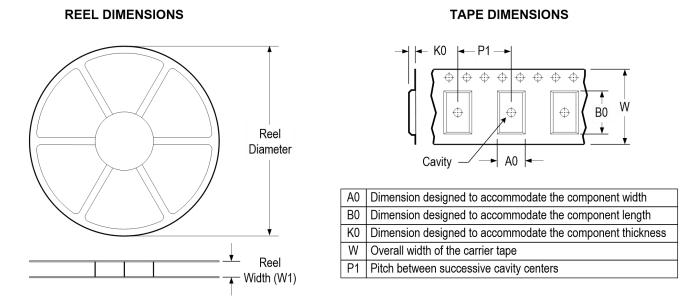


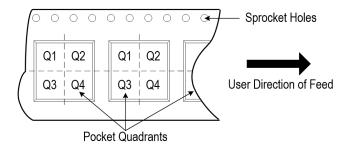
Figure 5. Differential Amplifier



## Tape and Reel Information



#### **QUADRANT ASSIGNMENTS FOR PIN 1 ORIETATION IN TAPE**



#### \* All dimensions are nominal

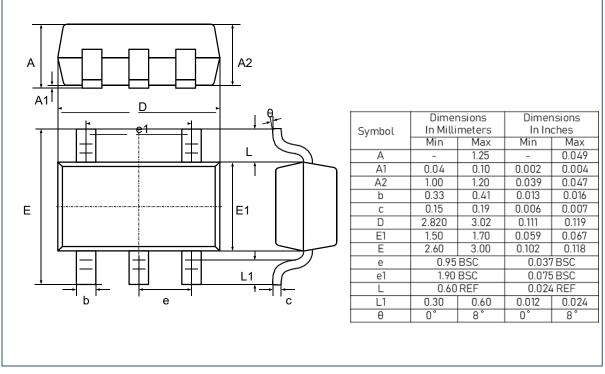
Device	Package Type	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin 1 Quadrant
LTA8051XT5/R6	SOT23	5	3 000	178	9.0	3.3	3.2	1.5	4.0	8.0	Q3
LTA8051XS8/R8	SOIC	8	4 000	330	12.4	6.6	5.3	2.0	8.0	12.0	Q1
LTA8052XS8/R8	SOIC	8	4 000	330	12.4	6.6	5.3	2.0	8.0	12.0	Q1
LTA8052XV8/R6	MSOP	8	3 000	330	12.4	5.0	3.5	2.0	8.0	12.0	Q1
LTA8054XS14/R5	SOIC	14	2 500	330	12.4	6.5	9.5	2.0	8.0	16.0	Q1
LTA8054XT14/R6	TSSOP	14	3 000	330	12.4	6.9	5.5	1.2	8.0	16.0	Q1

### P-17

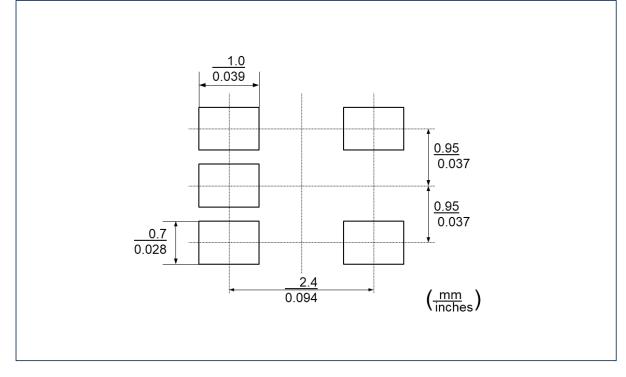
# LTA8051, LTA8052, LTA8054 200 MHz, High Slew Rate, RRO, CMOS Amplifiers

# Package Outlines

## DIMENSIONS, SOT23-5L



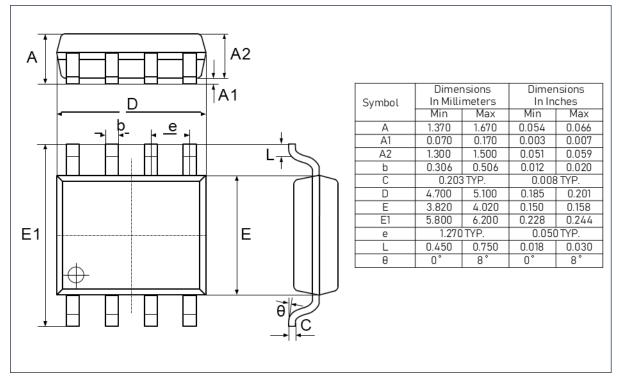
### RECOMMENDED SOLDERING FOOTPRINT, SOT23-5L



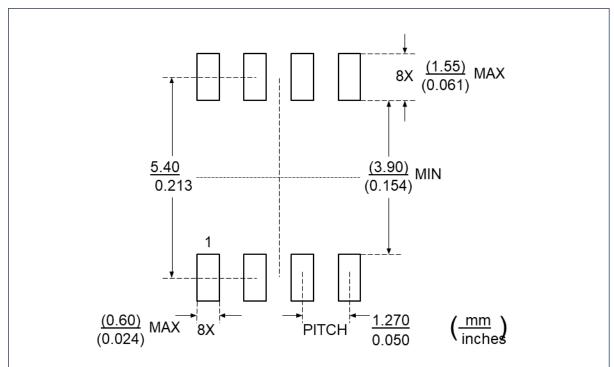


# Package Outlines (cont.)

#### DIMENSIONS, SOIC-8L



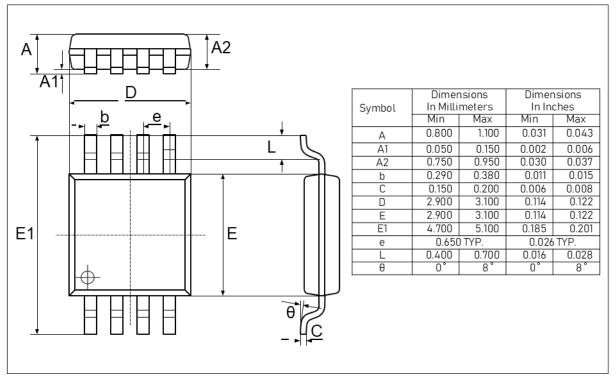
#### RECOMMENDED SOLDERING FOOTPRINT, SOIC-8L



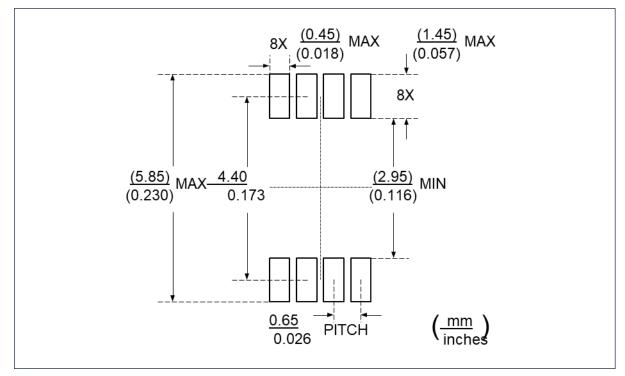


# Package Outlines (cont.)

#### DIMENSIONS, MSOP-8L



### RECOMMENDED SOLDERING FOOTPRINT, MSOP-8L



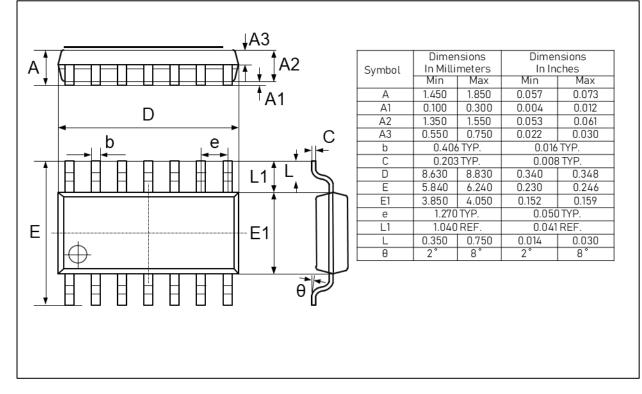


#### P-20

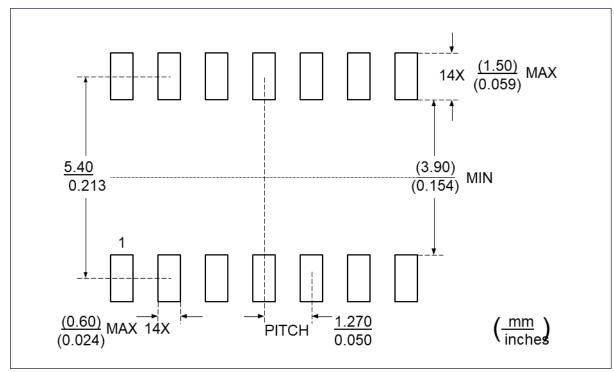
# LTA8051, LTA8052, LTA8054 200 MHz, High Slew Rate, RRO, CMOS Amplifiers

# Package Outlines (Cont.)

### DIMENSIONS, SOIC-14L



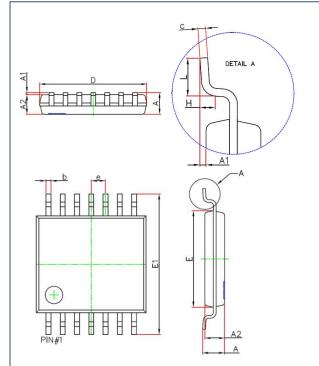
#### RECOMMENDED SOLDERING FOOTPRINT, SOIC-14L





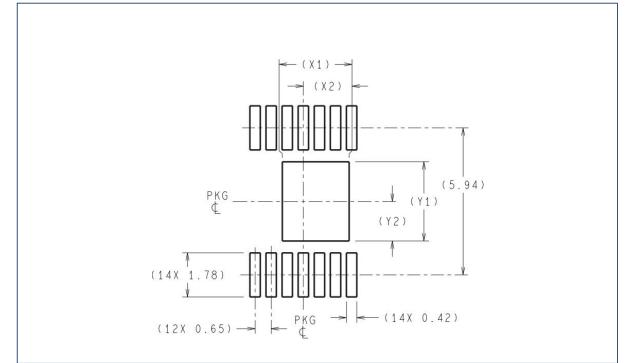
# Package Outlines (Cont.)

#### DIMENSIONS, TSSOP-14L



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min.	Max.	Min.	Max.
D	4.9	5.1	0.193	0.201
E	4.3	4.5	0.169	0.177
b	0.19	0.3	0.007	0.012
С	0.09	0.2	0.004	0.008
E1	6.25	6.55	0.246	0.258
Α	_	1.2	_	0.047
A2	0.8	1	0.031	0.039
A1	0.05	0.15	0.002	0.006
е	0.65 (BSC)		0.026 (BSC)	
L	0.5	0.7	0.02	0.028
Н	0.25(TYP)		0.01(TYP)	
θ	1°	7°	1°	7°

#### RECOMMENDED SOLDERING FOOTPRINT, TSSOP-14L





#### **Important Notice**

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